PID controllers

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PID Control PID parameters

A survey of controllers for more than 100 boiler-turbine units in the Guangdong Province in China [1]:

- 94.4% of all controllers were PI,
- 3.7% PID,
- 1.9% used advanced control.



$$u(t) = k_{p} \cdot e(t) + k_{i} \int_{0}^{t} e(\tau) d\tau + k_{d} \frac{de(t)}{dt} = k_{p} \left(e(t) + \frac{1}{T_{i}} \int_{0}^{t} e(\tau) d\tau + T_{d} \frac{de(t)}{dt} \right), \quad (1)$$

$$k_{i} = \frac{k_{p}}{T_{i}}, \quad (2)$$

$$k_{d} = k_{p} \cdot T_{d}. \quad (3)$$

[1] Li Sun, Donghai Li, and Kwang Y. Lee. Optimal disturbance rejection for pi controllerwith constraints on relative delay margin.ISA Transactions,2016.

PID Control PID parameters



Figure 11.2: Responses to step changes in the reference value for a system with a proportional controller (a). PI controller (b) and PID controller (c). The process has the transfer function $P(s) = 1/(s+1)^3$, the proportional controller has parameters $k_p = 1$, 2 and 5, the PI controller has parameters $k_p = 1$, $k_i = 0$, 0.2, 0.5, and 1, and the PID controller has parameters $k_p = 2.5$, $k_i = 1.5$ and $k_d = 0$, 1, 2, and 4.

$$u = k_{p} \cdot e + k_{i} \int_{0}^{t} e(\tau) d\tau + k_{d} \frac{de}{dt} = k_{p} \left(e + \frac{1}{T_{i}} \int_{0}^{t} e(\tau) d\tau + T_{d} \frac{de}{dt} \right)$$
(4)

- If the plant mathematical model cannot be obtained at all, then an analytical or computational approach to the design of a PID controller is not possible.
- Then we must resort to experimental approaches to the tuning of PID controllers.
- Ziegler and Nichols suggested rules for tuning PID controllers (values K_p , T_i , and T_d) based on:
 - Experimental step responses (Method 1).
 - Based on the value of *K_p* that results in marginal stability when only proportional control action is used (Method 2).



- We obtain experimentally the response of the plant to a unit-step input.
- This method applies if the response to a step input exhibits an S-shaped curve.
- If the plant involves neither integrator(s) nor dominant complex-conjugate poles, then such a unit-step response curve may look S-shaped.



• The S-shaped curve may be characterized by two constants, delay time *L* and time constant *T*.



 Table 8-1
 Ziegler-Nichols Tuning Rule Based on Step Response of Plant (First Method)

| Type of Controller | K_p | T_i | T_d |
|-----------------------|-------------------|-----------------|-------|
| Р | $\frac{T}{L}$ | ∞ | 0 |
| PI | $0.9 \frac{T}{L}$ | $\frac{L}{0.3}$ | 0 |
| PID | $1.2 \frac{T}{L}$ | 2L | 0.5L |

function C(s)/U(s) may then be approximated by a first-order system with a transport lag as follows:

$$\frac{C(s)}{U(s)} = \frac{Ke^{-Ls}}{Ts+1}$$

Ziegler and Nichols suggested to set the values of K_p , T_i , and T_d according to the formula shown in Table 8–1.

Notice that the PID controller tuned by the first method of Ziegler-Nichols rules gives

$$\begin{split} G_c(s) &= K_p \bigg(1 + \frac{1}{T_i s} + T_d s \bigg) \\ &= 1.2 \, \frac{T}{L} \left(1 + \frac{1}{2Ls} + 0.5Ls \right) \\ &= 0.6T \, \frac{\bigg(s + \frac{1}{L} \bigg)^2}{s} \end{split}$$

Thus, the PID controller has a pole at the origin and double zeros at s = -1/L.

- We first set $T_i = \infty$, and $T_d = 0$.
- Increase K_p from 0 to a critical value K_{cr} at which the output first exhibits sustained oscillations.
- Thus, the critical gain K_{cr} and the corresponding period P_{cr} are experimentally determined.
- If the output does not exhibit sustained oscillations, then this method does not apply.



Table 8-2Ziegler-Nichols Tuning Rule Based on Critical Gain
 $K_{\rm cr}$ and Critical Period $P_{\rm cr}$ (Second Method)

| Type of Controller | K_p | T_i | T_d |
|-----------------------|---------------------|----------------------------|----------------------|
| Р | $0.5K_{\rm cr}$ | ∞ | 0 |
| PI | 0.45K _{cr} | $\frac{1}{1.2} P_{\rm cr}$ | 0 |
| PID | $0.6K_{\rm cr}$ | $0.5P_{\rm cr}$ | 0.125P _{cr} |

Notice that the PID controller tuned by the second method of Ziegler-Nichols rules gives

$$\begin{aligned} G_c(s) &= K_p \bigg(1 + \frac{1}{T_i s} + T_d s \bigg) \\ &= 0.6 K_{\rm cr} \bigg(1 + \frac{1}{0.5 P_{\rm cr} s} + 0.125 P_{\rm cr} s \bigg) \\ &= 0.075 K_{\rm cr} P_{\rm cr} \frac{\bigg(s + \frac{4}{P_{\rm cr}} \bigg)^2}{s} \end{aligned}$$

Thus, the PID controller has a pole at the origin and double zeros at $s = -4/P_{cr}$.

EXAMPLE 8–1 Consider the control system shown in Figure 8–6 in which a PID controller is used to control the system. The PID controller has the transfer function

$$G_c(s) = K_p \left(1 + \frac{1}{T_i s} + T_d s \right)$$

Although many analytical methods are available for the design of a PID controller for the present system, let us apply a Ziegler–Nichols tuning rule for the determination of the values of parameters K_p , T_i , and T_d . Then obtain a unit-step response curve and check to see if the designed system exhibits approximately 25% maximum overshoot. If the maximum overshoot is excessive (40% or more), make a fine tuning and reduce the amount of the maximum overshoot to approximately 25% or less.



$$K_{cr} = 30$$

$$P_{cr} = 2.8099$$

$$K_{p} = 0.6 \cdot K_{cr} = 18$$

$$T_{i} = 0.5 \cdot P_{cr} = 1.405$$

$$T_{d} = 0.125 \cdot P_{cr} = 0.35124$$



PID Control Tuning Method 2, Example 8-1, III

Maximum overshoot is close to 62%.

| MATLAB Program 8–1 | | |
|---|--|--|
| % Unit-step response | | |
| num = [6.3223 18 12.811]; den = [1 6 11.3223 18 12.811]; step(num,den) grid title('Unit-Step Response') | | |



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PID Control Tuning Method 2, Example 8-1, IV

The Ziegler–Nichols tuning rule has provided a starting point for fine tuning.

 $K_p = 39.42$ $T_i = 3.077$ $T_d = 0.7692$

Maximum overshoot is fairly close to 25%.



- The value of Kp increases the speed of response.
- However, varying the location of the double zero has a significant effect on the maximum overshoot.

- Karl J. Astrom and Richard M. Murray *Feedback Systems*. Version v3.0i. Princeton University Press. September 2018. Chapter 11.
- Ogata, Katsuhiko. Modern Control Engineering. Fifth Edition. Prentice Hall. 2009. Chapter 8.